

## B.E.

Sixth Semester Examination, Dec.-2007

### Automatic Controls (ME-308-E)

**Note :** Attempt any *five* questions.

**Q. 1.** Draw root loci for a system with  $GH(s) = K / [s(s+2)(s+3)]$  and find its intersect on the imaginary axis. Also find the value of  $K$  for which this system will be unstable.

**Ans.** We have,  $G(s)H(s) = \frac{K}{s(s+2)(s+3)}$

1. The finite poles of open loop transfer function are at  $s = 0$ ,  $s = -2$  and  $s = -3$ .
2. As the system has no zero, so the root loci would terminate at the zeroes located at infinity.
3. The number of root loci,  $P = 3$ .
4. The given transfer function is rational, so the root loci is symmetrical about real axis.
5. The angles of asymptotes for  $K > 0$  are :

$$K = 0, \quad \theta_1 = \frac{\pi}{n-m} = \frac{\pi}{3-0} = \frac{\pi}{3} = 60^\circ$$

$$K = 1, \quad \theta_2 = \frac{3\pi}{3} = 180^\circ$$

$$K = 2, \quad \theta_3 = \frac{5\pi}{3} = 300^\circ$$

6. Intercepts of asymptotes at real axis,

$$\begin{aligned} &= \frac{(P_1 + P_2 + P_3 + \dots + P_n) - (Z_1 + Z_2 + Z_3 + \dots + Z_m)}{n-m} \\ &= \frac{(0 - 2 - 3) - 0}{3-0} = \frac{-5}{3} = -1.66 \end{aligned}$$

7. Root locus on real axis :

(a) Between  $s = 0$  to  $s = -2$  and

(b) Between  $s = -3$  to  $-\infty$ .

8. Intersection with imaginary axis :

The characteristic equation is,

$$s(s+2)(s+3) + K = 0$$

$$\Rightarrow s^3 + 5s^2 + 6s + K = 0$$

Put  $s = j\omega$ , we have

$$-j\omega^3 - 5\omega^2 + 6j\omega + K = 0$$

Equating imaginary parts equal to zero,

$$-j\omega^3 + 6j\omega = 0$$

$$\therefore \omega = \pm\sqrt{6} \approx \pm 2.45$$

Now, equating real parts equal to zero and putting the value of  $\omega = \pm\sqrt{6}$ , gives

$$K = 5 \times 6 = 30$$

So the loci intersect the imaginary axis at  $\pm j\sqrt{6}$  with  $K = 30$ .

9. Break away point : The characteristic equation is,

$$s^3 + 5s^2 + 6s + K = 0$$

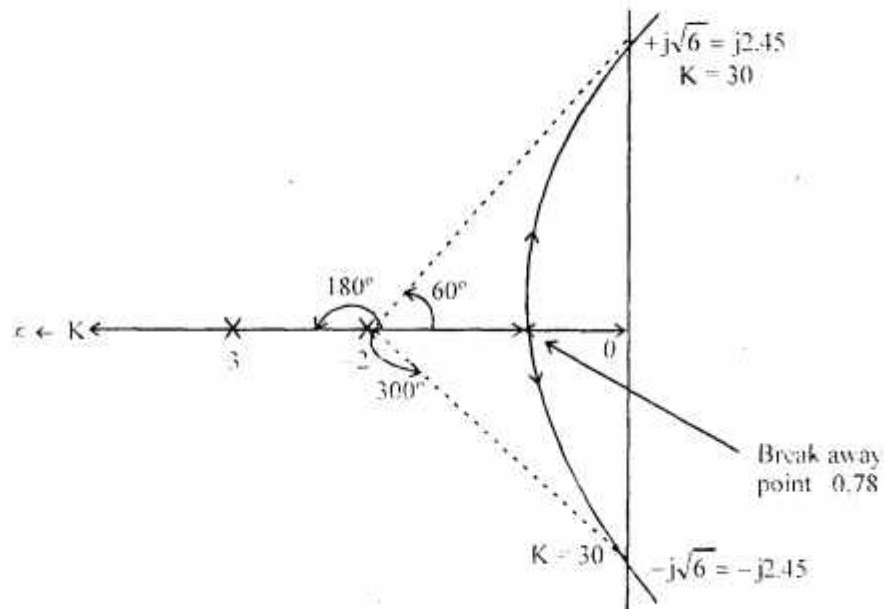
$$\therefore K = -(s^3 + 5s^2 + 6s)$$

$$\text{So, } \frac{dK}{ds} = -(3s^2 + 10s + 6) = 0$$

$$\begin{aligned} \text{Hence, } s &= \frac{-10 \pm \sqrt{100 - 72}}{6} \\ &= -2.55 \text{ and } -0.78 \end{aligned}$$

The value -2.55 is not lying on root locus. So the valid value is  $-0.78$ .

The root loci are given below :

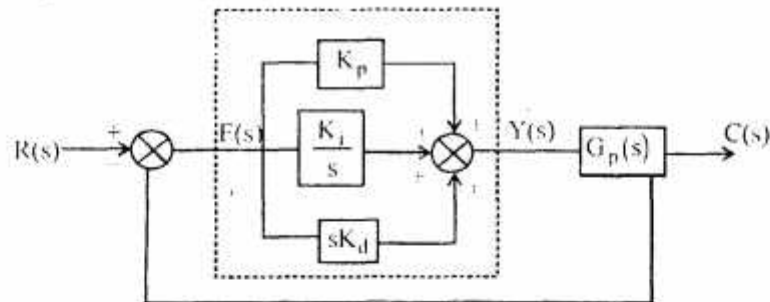


Value of K for unstable system.

Since the root locus intersects the imaginary axis at  $K = 30$ . So for  $K > 30$ , the root locus lies in the right half of s-plane. So the system is unstable for  $K > 30$ .

**Q. 2. Derive an expression for transfer function of a PDI type hydraulic controller.**

**Ans.** The block diagram of a PDI type hydraulic controller is shown below :



In the diagram, the PDI type hydraulic controller is shown inside the dotted lines.

Let,  $K_p$  be the proportional gain of the controller

$K_i$  be the integral gain of the controller

$K_d$  be the derivative gain of the controller

and  $G_p(s)$  be the transfer function of the system, whose output is to be controlled by the PDI controller

$E(s)$  be input signal to the controller or actuating signal

$Y(s)$  be the output of the controller

As from the block diagram, the output of the controller is given by (in time domain)

$$Y(t) = K_p e(t) + K_i \int_0^t e(t) dt + K_d \frac{de(t)}{dt}$$

Taking Laplace transformation, we have

$$Y(s) = K_p E(s) + \frac{K_i}{s} E(s) + sK_d E(s)$$

So, the transfer function of the controller is

$$\begin{aligned} \frac{Y(s)}{E(s)} &= K_p + \frac{K_i}{s} + sK_d \\ &= K_p \left[ 1 + \frac{K_i}{sK_p} + \frac{sK_d}{K_p} \right] \end{aligned}$$

Or

$$\frac{Y(s)}{E(s)} = K_p \left[ 1 + \frac{1}{sK_p / K_i} + \frac{sK_d}{K_p} \right]$$

If we write

$$\frac{K_p}{K_i} = T_i$$

and

$$\frac{K_d}{K_p} = T_d$$

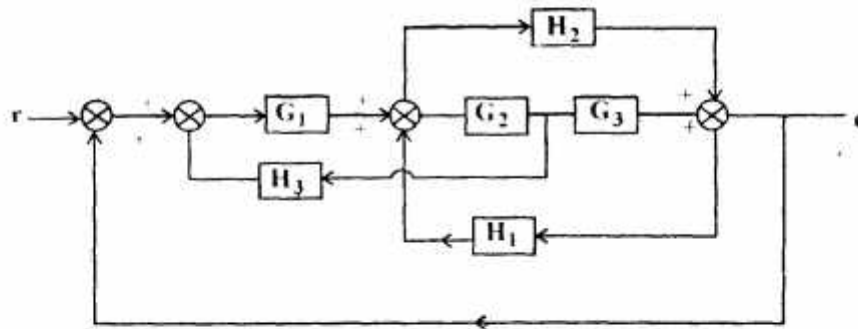
Then the transfer function of a PDI type hydraulic controller is given by

$$G_c = K_p \left[ 1 + \frac{1}{sT_i} + sT_d \right]$$

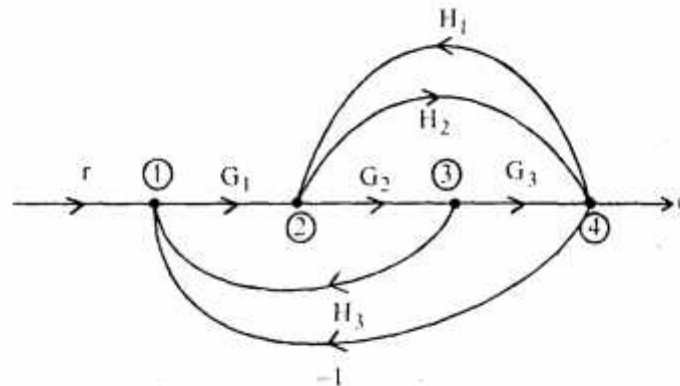
Where,  $T_i = \frac{K_p}{K_i}$  is called integral time constant of the PDI controller.

and  $T_d = \frac{K_d}{K_p}$  is called derivative time constant of the PDI controller.

**Q. 3. For the block diagram in figure, draw signal flow diagrams and derive expression for  $c/r$ , using Mason's formula. Check by block diagram algebra.**



**Ans.** The signal flow graph of the given block diagram is given below :

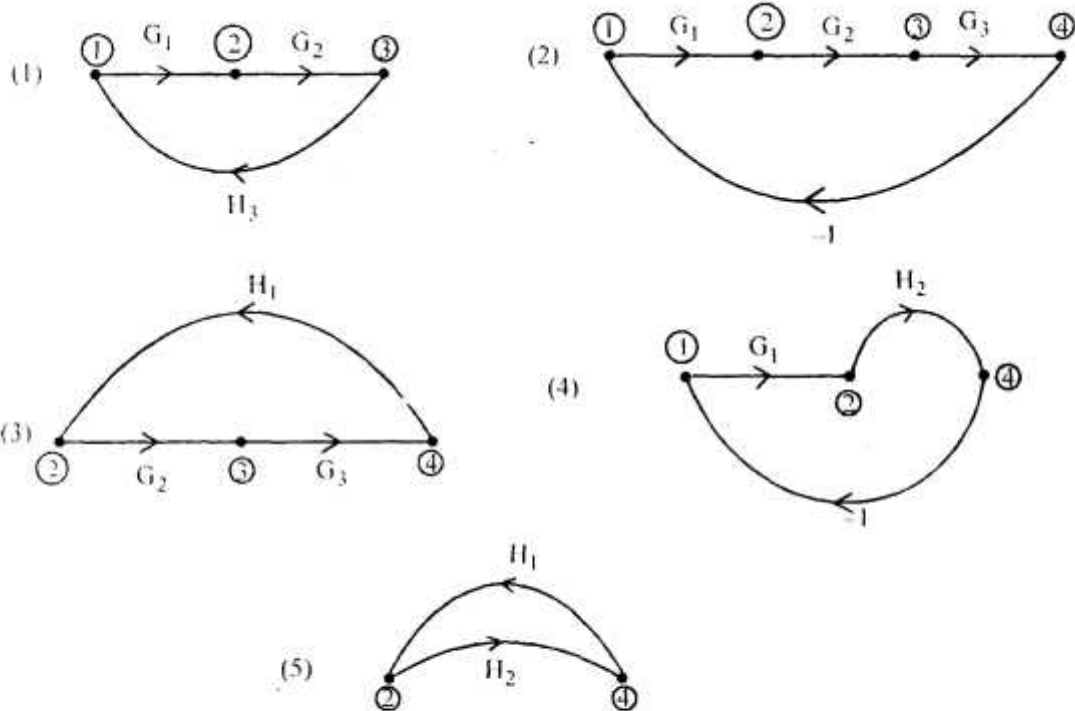


According to the Mason's formula :

$$\begin{aligned} \frac{c}{r} &= \frac{1}{\Delta} \sum_{k=1}^p M_k \Delta_k \\ &= \frac{1}{\Delta} (M_1 \Delta_1 + M_2 \Delta_2 + \dots + M_p \Delta_p) \end{aligned}$$

$\Delta = 1 - (\text{Sum of all individual loops}) + \text{sum of the products of loop gains of all possible combination of non-touching loops taken two at a time}.$

There are 5 individual loops :



The sum of gains of these 5 loops,

$$= G_1 G_2 H_3 - G_1 G_2 G_3 + G_2 G_3 H_1 - G_1 H_2 + H_1 H_2$$

So,

$$\Delta = 1 - G_1 G_2 H_3 + G_1 G_2 G_3 - G_2 G_3 H_1 + G_1 H_2 - H_1 H_2$$

There are 2 Nos. of forward paths,



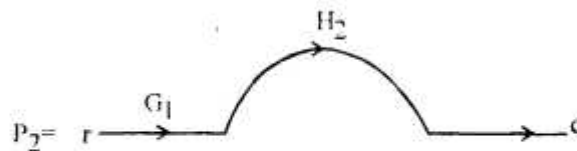
Path gain,

$$M_1 = G_1 G_2 G_3$$

There is no non-touching loop

So,

$$\Delta_1 = 1$$



Path gain

$$= G_1 H_2 = M_2$$

There is no non-touching loop. So,

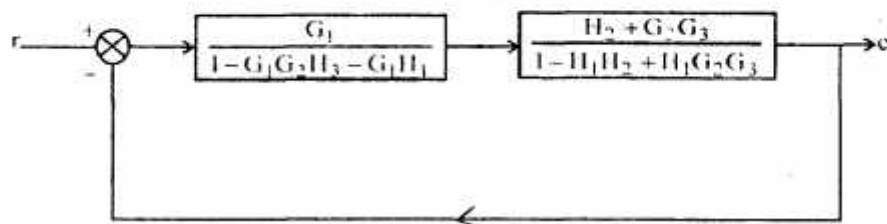
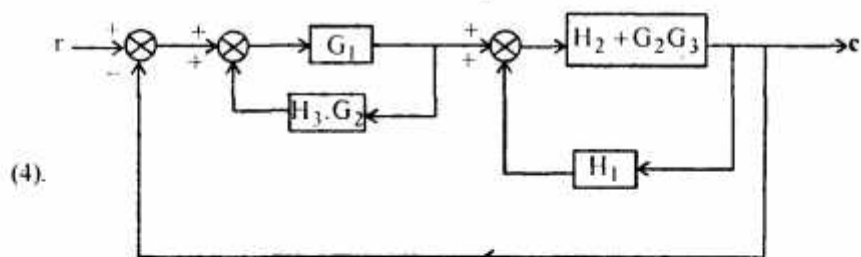
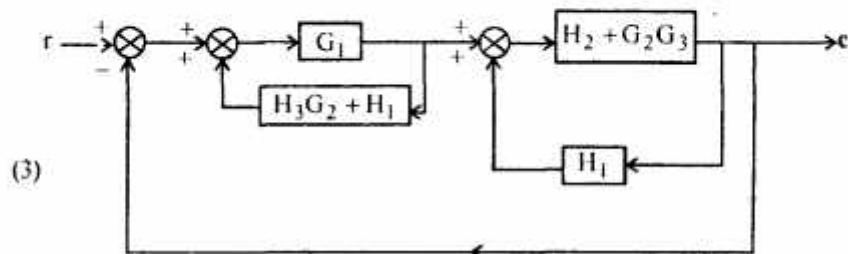
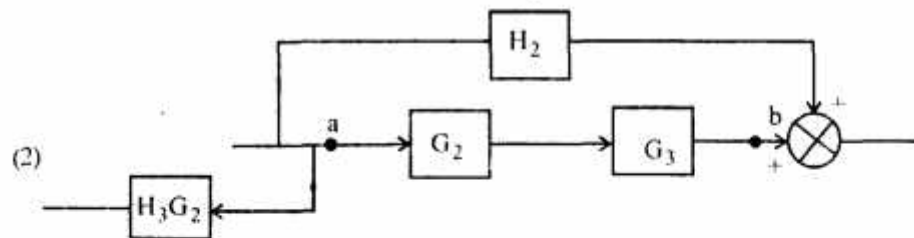
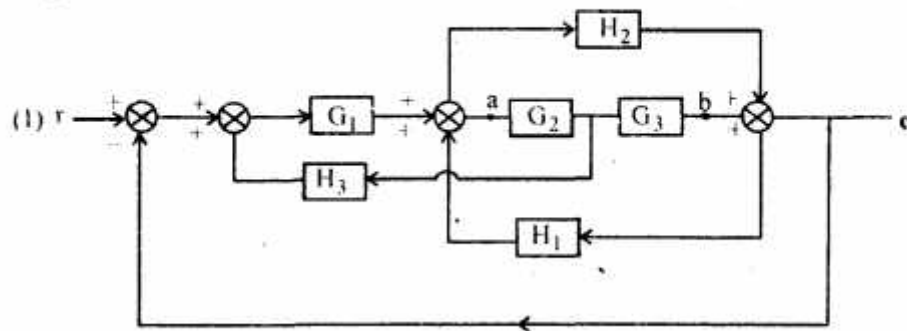
$$\Delta_2 = 1$$

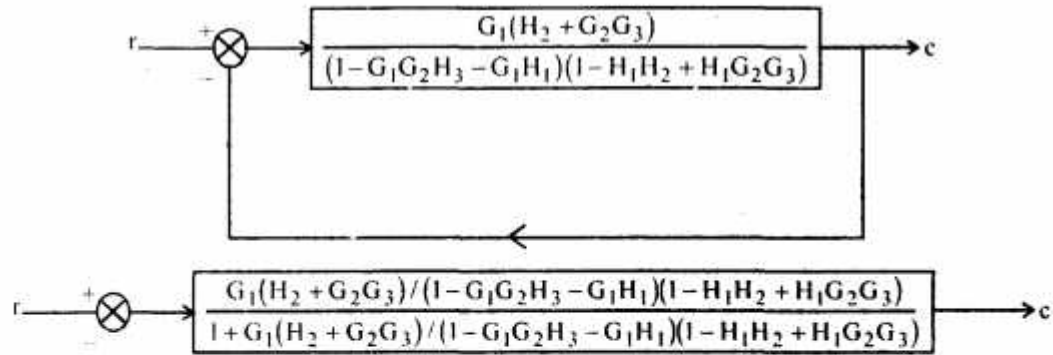
Hence,

$$\frac{c}{r} = \frac{1}{\Delta} (M_1 \Delta_1 + M_2 \Delta_2)$$

$$= \frac{G_1 G_2 G_3 + G_1 H_2}{1 - G_1 G_2 H_3 + G_1 G_2 G_3 - G_2 G_3 H_1 + G_1 H_2 - H_1 H_2}$$

**Block diagram reduction :** The successively reduced block diagrams are given below :



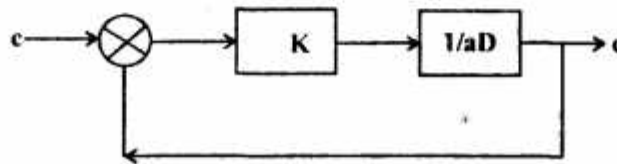


So, on solving we have

$$\frac{c}{r} = \frac{G_1 G_2 G_3 + G_1 H_2}{1 - G_1 G_2 H_3 + G_1 G_2 G_3 - G_2 G_3 H_1 + G_1 H_2 - H_1 H_2}$$

which is same as found by Mason's formula.

**Q. 4. Draw polar plots of the following first order system :**



**Ans.** The open loop transfer function of the given first order system is given by :

$$G_o(t) = \frac{K}{sD}$$

Taking Laplace transform, we have

$$G_o(s) = \frac{K}{as}$$

The closed loop transfer function is :

$$G(s) = \frac{K}{sa + K}$$

Putting  $s = j\omega$ , we have

$$G(j\omega) = \frac{K}{j\omega a + K} = \frac{K}{K^2 + a^2 \omega^2} - j \frac{a\omega K}{K^2 + a^2 \omega^2}$$

Now,  $|G(j\omega)| = \frac{K}{\sqrt{K^2 + a^2 \omega^2}} \quad \dots(1)$

and  $\phi = -\tan^{-1} \frac{a\omega}{K} \quad \dots(2)$



(i) Now, when  $\omega = 0$ ,  $|G(j\omega)| = \frac{K}{\sqrt{K^2}} = 1$

and  $\phi = -\tan^{-1} 0 = 0$

It is also the intersecting point with the real axis of the plot.

(ii) When  $\omega$  tends  $\infty$ , then

$$|G(j\omega)| = 0$$

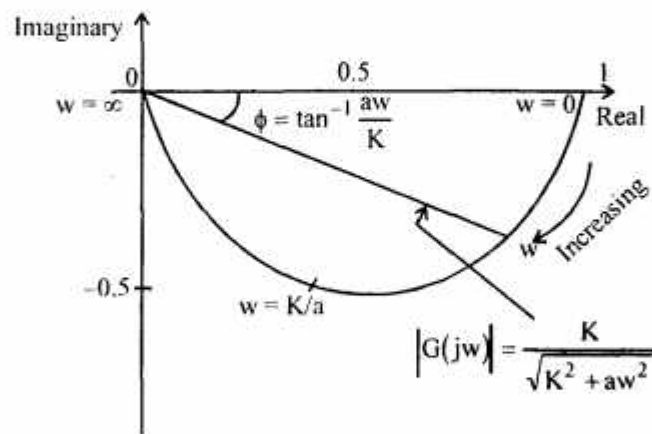
and  $\phi = -90^\circ$

This is the point where the plot crosses the imaginary axis.

(iii) When  $\omega = K/a$

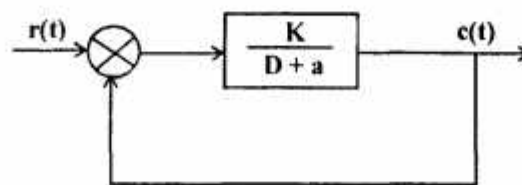
$$|G(j\omega)| = \frac{K}{\sqrt{K^2 + K^2}} = \frac{1}{\sqrt{2}}$$

$$\phi = -\tan^{-1} 1 = -45^\circ$$



The polar plot is drawn in the figure shown above.

**Q. 5. For the first order system shown in figure, derive the solution for output  $C(t)$  as a function of time a unit step input  $r(t) = 1$ , using time domain analysis.**



**Ans.**

The open loop transfer function of the given system is  $\frac{K}{D+a}$ .

The closed loop transfer function of this system in  $s$  domain is given by,



$$G(s) = \frac{C(s)}{R(s)} = \frac{K}{s+a+K+1}$$

$$= \frac{K}{(1+a+K)+s}$$

Hence,  $C(s) = \frac{K}{(1+a+K)} \cdot R(s)$  ... (1)

But  $r(t) = 1$  given

So,  $R(s) = \frac{1}{s}$  (taking Laplace transformation)

Putting this value in equation (1), we have,

$$C(s) = \frac{K}{(1+a+K)} \times \frac{1}{s}$$

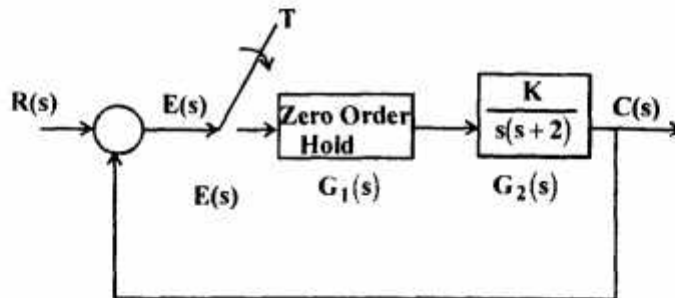
$$= \frac{K}{s[(1+a+K)+s]}$$

Taking inverse Laplace transformation, we have

$$C(t) = K \left[ 1 - e^{-(1+a+K)t} \right]$$

is the required solution of the given system.

**Q. 6. Find, using Routh's criterion if the system is stable or not. Take  $G_2(s) = K / [s(s+1)]$ ,  $T = 1$  sec and  $K = 1$ . If  $K = 10$ , check for stability.**



**Ans.** The transfer function of the given system is,

$$G(s) = \frac{K}{s(s+1)} \times \frac{K}{s(s+2)}$$

$$= \frac{K^2}{s^2(s+1)(s+2)} \times \frac{K^2}{s^4 + 3s^3 + 2s^2}$$

The characteristic equation is given by,

$$s^4 + 3s^3 + 2s^2 + K^2 = 0$$

(i) If  $K = 1$ , the characteristic equation is,

$$s^4 + 3s^3 + 2s^2 + 1 = 0$$

Here, as the coefficient of 's' is zero, so the system either unstable or critically stable.

The Routh's array is given by :

|       |               |   |
|-------|---------------|---|
| $s^4$ | 1             | 2 |
| $s^3$ | 3             | 1 |
| $s^2$ | $\frac{4}{3}$ | 0 |
| $s$   | 1             | 0 |
| $s^0$ | 0             | 0 |

There is no change of sign. So no pole lies in right half of s-plane, but as the coefficient of term 's' is missing. So the system is critically stable for  $K = 1$ .

(ii) Now,  $K = 10$

The Routh's array is given below :

|       |                 |     |
|-------|-----------------|-----|
| $s^4$ | 1               | 2   |
| $s^3$ | 3               | 100 |
| $s^2$ | $-\frac{94}{3}$ | 0   |
| $s^1$ | 100             | 0   |
| $s^0$ | 0               | 0   |

As the sign changes two times. So two poles lies in the right half of s-plane. So the system is unstable for  $K = 10$ .

**Q. 7. For the system with transfer function  $\frac{Y(s)}{U(s)} = \frac{s^2 + 2s + 1}{s^3 + 7s^2 + 14s + 8}$ . Derive the state-space representation.**

**Ans.** The transfer function of the given system is,

$$\frac{Y(s)}{U(s)} = \frac{s^2 + 2s + 1}{s^3 + 7s^2 + 14s + 8}$$

Breaking the T.F. into two parts, we have

$$\frac{Y(s)}{U(s)} = \frac{X_1(s)}{U(s)} \times \frac{Y(s)}{X_1(s)} = \frac{1}{s^3 + 7s^2 + 14s + 8} \times (s^2 + 2s + 1)$$

Now consider,

$$\frac{X_1(s)}{U(s)} = \frac{1}{s^3 + 7s^2 + 14s + 8}$$

$$\Rightarrow (s^3 + 7s^2 + 14s + 8)X_1(s) = U(s)$$

Taking inverse Laplace transformation, we have

$$\frac{d^3 x_1}{dt^3} + 7 \frac{d^2 x_1}{dt^2} + 14 \frac{dx_1}{dt} + 8x_1 = u(t)$$

$$\Rightarrow \ddot{x}_1(t) + 7\ddot{x}_1(t) + 14\dot{x}_1(t) + 8x_1(t) = u(t)$$

Hence,  $\ddot{x}_1(t) = -7\ddot{x}_1(t) - 14\dot{x}_1(t) - 8x_1(t) + u(t)$

Now select the state variables,

$$\dot{x}_1 = x_2$$

$$\ddot{x}_1 = \dot{x}_2 = x_3$$

$$\ddot{x}_1 = \dot{x}_3$$

Now,  $\dot{x}_3 = \ddot{x}_1 = -7\ddot{x}_1(t) - 14\dot{x}_1(t) - 8x_1(t) + u(t)$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -7 & -14 & 8 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} u(t)$$

Now consider,  $\frac{Y(s)}{X_1(s)} = (s^2 + 2s + 1)$

$$\Rightarrow Y(s) = (s^2 + 2s + 1)X_1(s)$$

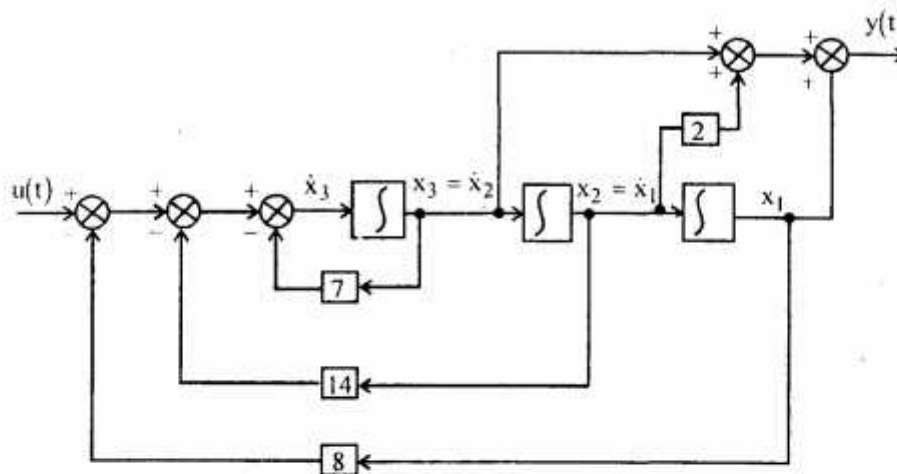
Taking inverse Laplace transformation, we have

$$\begin{aligned} y(t) &= \frac{d^2 x_1}{dt^2} + 2 \frac{dx_1}{dt} + x_1 \\ &= \ddot{x}_1(t) + 2\dot{x}_1(t) + x_1(t) \\ &= x_3 + 2\dot{x}_2 + x_1 \end{aligned}$$

Hence,  $y(t) = [1, 2, 1]x(t)$

$$= [1, 2, 1] \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}$$

The block diagram is given below :



**Q. 8. For a system with the characteristic equation :**

**$s^3 - 4s^2 + s + 6 = 0$ , find the number of roots, if any with positive real parts.**

**Ans.** The given characteristic equation is,

$$s^3 - 4s^2 + s + 6 = 0$$

The roots of this characteristic equation with positive real parts can be found by Routh's Hurwitz criterion as follows :

The Routh's array is given below :

|       |                |   |
|-------|----------------|---|
| $s^3$ | 1              | 1 |
| $s^2$ | -4             | 6 |
| $s^1$ | $-\frac{5}{2}$ | 0 |
| $s^0$ | 6              | 0 |

In the first column of the Routh array the sign changes two times. So, there are two roots with positive real parts.